

Diskretna Fourier-ova Transformacija - DFT

Zadat je $n \in \mathbb{N}$ i konačni uiz ili vektor $a \in \mathbb{C}^n$ od n elemenata

$$a = (a_0, a_1, \dots, a_{n-1})^T.$$

Taj uiz/vektor možemo interpretirati i kao funkciju

$$a : \mathbb{Z}_n \rightarrow \mathbb{C}$$

gdje je $\mathbb{Z}_n = \{0, 1, \dots, n-1\}$ standardni sustav ostataka modulo n .

Elemente vektora a indeksiraju elementima iz \mathbb{Z}_n , a ne iz $\{1, \dots, n\}$, kao što je uobičajeno u češćim algebrama. Razlog nije samo tradicija u obradi signala, nego i matematičku pozadinu.

Vektor a možemo zapisati kao diskretni uzorak nekog (može i kontinuiranog) signala, koji je "snimljen" u trenucima $t = 0, 1, \dots, n-1$.

Vektoru a možemo pripojiti polinom A nad \mathbb{C} , oblika

$$A(x) = \sum_{j=0}^{n-1} a_j x^j.$$

Takvi polinomi (kao i vektor a , duljine n) čine vektorski prostor nad \mathbb{C} , dimenzije n , izomorfan \mathbb{C}^n . Zato kažemo da polinom A ima red n , f.j. stupanj je $\leq n-1$.

Vektore, gledane kao funkcije na \mathbb{Z}_n , možemo i množiti - po fočkama. Analogno, polinome možemo množiti (algebra!). Međutim, odmah uočavamo da ovakvo pripajanje

$$a \mapsto A$$

može homomorfizam obzirom na množenje (Očito je homomorfizam na ostale operacije - zbrajanje množenje skalarom - f.j. između vektorskih prostora).

Naine, za "funkcijsko" ili Hadamard-ovo množenje vektora mijedi

$$c = a \cdot b \Leftrightarrow c_j = a_j \cdot b_j, j=0, \dots, n-1,$$

dok za množenje polinoma mijedi

$$C = A \cdot B \Leftrightarrow c_j = \sum_{k=0}^j a_k \cdot b_{j-k}, j=0, \dots, 2n-1$$

pa je produkt C reda $2n-1$.

Da bismo uspostavili vezu s produktem polinoma, na vektorsima uvođimo novu operaciju \otimes , kogu zovemo konvolucija, tako da je

$$(a \otimes b)_j = \sum_{k=0}^{n-1} a_k b_{j-k}, j=0, \dots, n-1.$$

\hookrightarrow periodični pros.

Ovo još niječ ne odgovara relaciji za koeficijente produkta polinoma. I ne može, jer redovi (dužine) nisu ujednačeni.

Međutim, ako uizove/vektore a i b produljimo nulanom do dovoljne dužine N, pripadaju polinomu A i B ostaju isti, samo ih gledamo u većem vektorskome prostoru. Naravno, i C=A·B se ne mijenja. Zbog toga je dovoljno uzeti $N \geq 2n-1$, pa da i niz c koeficijenata produkta bule cijeli prikaziv.

Dakle, napravimo

$$\begin{aligned} a &\mapsto a' = (a_0, \dots, a_{n-1}, 0, \dots, 0)^T \\ b &\mapsto b' = (b_0, \dots, b_{n-1}, 0, \dots, 0)^T \end{aligned} \quad \} \text{dužine } N$$

$$\therefore c' = a' \otimes b' = (c'_0, \dots, c'_{N-1})$$

Ako je $N \geq 2n-1$, onda mijedi

$$c'_j = \sum_{k=0}^{N-1} a'_k b'_{j-k} = \sum_{k=0}^j a_k b_{j-k}, j=0, \dots, 2n-2$$

$$c'_j = 0, j > 2n-2 \quad (j=2n-1, \dots, N-1)$$

i stavimo:

$$c := (c'_0, \dots, c'_{2n-2})^T - \text{dužina } 2n-1.$$

Dobiveni už C je pribljeni polinomu $C = A \cdot B$.
 (Precizuje, dobivenom užu C pribljen je polinom $C = A \cdot B$).

Osim smo uspostavili formalni "homomorfizam"
 konvolucije užova i produkta polinoma. Stamo,
 još ušta konisno nismo uopravili, dok nemamo
efikasne algoritme za računanje nečeg od toga!

Polinome zasad pokazujemo koeficijentima - u tzv.
 koeficijentnoj reprezentaciji. Ako se opet slijedimo
 interpolacije, možemo konstrui i drugačiji prikaz -
 vrijednostima na dovoljno veliku skupu točaka.

Savim općenito, za polinom A reda n možemo
 odabrat: bilo koji skup od točno n različitih
 točaka $z_0, \dots, z_{n-1} \in \mathbb{C}$ i dobiti tzv. vrijednosnu
 reprezentaciju - vektorom vrijednosti:

$$(A(z_0), \dots, A(z_{n-1}))^T \in \mathbb{C}^n.$$

Uočimo da u ovakvoj reprezentaciji, sve aritmetičke
 operacije na polinomima imaju jednostavan
 oblik - po točkama. To ujedoli i za produkt!

$$(A \cdot B)(z) = A(z) \cdot B(z)$$

Ako pazimo na red, sve operacije su efikasne -
 brzo se izrade - linearno u duljini užova.

Ovo što nam fali je efikasan prelaz između
 te druge reprezentacije

vektor koef. \longleftrightarrow vektor vrijednosti.

Napomena: ovo ima smisla samo za produkt, ako
 nam uspije.

Zasto? Sve ostale operacije (zbajanje, množenje
 skalarom) idu brzo u obje reprezentacije - tj.
 i u koeficijentnoj. Međutim, za množenje imamo
 samo standardni $\Theta(n^2)$ algoritam u koeficijentnoj
 reprezentaciji; pa tu držimo ušteku!

Efikasni prelaz iz jedne u drugu reprezentaciju
(i nadrag) dobivamo pomešavim izborom točaka
 z_0, z_1, z_{n-1} .

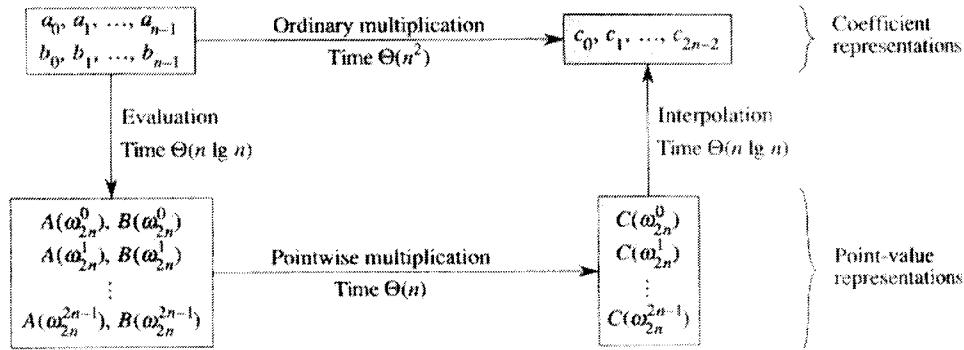


Figure 32.1 A graphical outline of an efficient polynomial-multiplication process. Representations on the top are in coefficient form, while those on the bottom are in point-value form. The arrows from left to right correspond to the multiplication operation. The ω_{2n} terms are complex $(2n)$ th roots of unity.

input and output representations are in coefficient form. We assume that n is a power of 2; this requirement can always be met by adding high-order zero coefficients.

1. *Double degree-bound:* Create coefficient representations of $A(x)$ and $B(x)$ as degree-bound $2n$ polynomials by adding n high-order zero coefficients to each.
2. *Evaluate:* Compute point-value representations of $A(x)$ and $B(x)$ of length $2n$ through two applications of the FFT of order $2n$. These representations contain the values of the two polynomials at the $(2n)$ th roots of unity.
3. *Pointwise multiply:* Compute a point-value representation for the polynomial $C(x) = A(x)B(x)$ by multiplying these values together pointwise. This representation contains the value of $C(x)$ at each $(2n)$ th root of unity.
4. *Interpolate:* Create the coefficient representation of the polynomial $C(x)$ through a single application of an FFT on $2n$ point-value pairs to compute the inverse DFT.

Steps (1) and (3) take time $\Theta(n)$, and steps (2) and (4) take time $\Theta(n \lg n)$. Thus, once we show how to use the FFT, we will have proven the following.

Theorem 32.2

The product of two polynomials of degree-bound n can be computed in time $\Theta(n \lg n)$, with both the input and output representations in coefficient form. ■

3. Konvolucijski teorema:

$$\text{DFT}_n(a \otimes b) = \text{DFT}_n(a) \cdot \text{DFT}_n(b)$$

(sto samo vec' konstrukci, zapravo, dozvati, ali za posebne mizove a' , b').

32.2-3

Do Exercise 32.1-1 by using the $\Theta(n \lg n)$ -time scheme.

32.2-4

Write pseudocode to compute DFT_n^{-1} in $\Theta(n \lg n)$ time.

32.2-5

Describe the generalization of the FFT procedure to the case in which n is a power of 3. Give a recurrence for the running time, and solve the recurrence.

32.2-6 *

Suppose that instead of performing an n -element FFT over the field of complex numbers (where n is even), we use the ring \mathbb{Z}_m of integers modulo m , where $m = 2^{tn/2} + 1$ and t is an arbitrary positive integer. Use $w = 2^t$ instead of ω_n as a principal n th root of unity, modulo m . Prove that the DFT and the inverse DFT are well defined in this system.

32.2-7

Given a list of values z_0, z_1, \dots, z_{n-1} (possibly with repetitions), show how to find the coefficients of the polynomial $P(x)$ of degree-bound n that has zeros only at z_0, z_1, \dots, z_{n-1} (possibly with repetitions). Your procedure should run in time $O(n \lg^2 n)$. (*Hint:* The polynomial $P(x)$ has a zero at z_j if and only if $P(x)$ is a multiple of $(x - z_j)$.)

32.2-8 *

$$y_k = \sum_{j=0}^{n-1} a_j z^{jk}$$

The **chirp transform** of a vector $a = (a_0, a_1, \dots, a_{n-1})$ is the vector $y = (y_0, y_1, \dots, y_{n-1})$, where $y_k = \sum_{j=0}^{n-1} a_j z^{jk}$ and z is any complex number. The DFT is therefore a special case of the chirp transform, obtained by taking $z = \omega_n$. Prove that the chirp transform can be evaluated in time $O(n \lg n)$ for any complex number z . (*Hint:* Use the equation

$$y_k = z^{k^2/2} \sum_{j=0}^{n-1} (a_j z^{j^2/2}) (z^{-(k-j)^2/2})$$

to view the chirp transform as a convolution.)

32.3 Efficient FFT implementations

Since the practical applications of the DFT, such as signal processing, demand the utmost speed, this section examines two efficient FFT implementations. First, we shall examine an iterative version of the FFT algorithm that runs in $\Theta(n \lg n)$ time but has a lower constant hidden in the Θ -notation than the recursive implementation in Section 32.2. Then, we shall use the insights that led us to the iterative implementation to design an efficient parallel FFT circuit.